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UNITED STATES PATENT APPLICATION

FOR

BURIED INTERSIGNAL CAPACITANCE

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BACKGROUND

1. Field of the Invention

This invention relates to printed circuit board (PCB) design. In particular, the invention relates to capacitance fabricated in the printed circuit board.

2. Description of Related Art

Mode compensation is one known signal quality technique in microwave PCB design. It is used to compensate for the different odd and even mode wave velocities through microstrip, where fields through air and dielectric cause the odd mode to be faster. The microwave application is for essentially analog signals and allows, for example, near-ideal directional couplers to be fabricated. In the present work, the mode compensation is used to help minimize the upward glitch seen on system memory data lines (i.e., when one or more lines are held low and the other data lines switch from low to high).

Mode compensation requires placing capacitance between adjacent signal traces, usually at the driver or receiver end of the trace. One way to implement mode compensation using the existing technology is to place discrete capacitors on the PCB, with highest priority to the receiver end of the signal traces. This method, however, is difficult and expensive. Placing discrete capacitors at the receiver end requires adding many discrete components, which add to the bill-of-materials (BOM) cost of the PCB. Furthermore, laying out a PCB to add sites to place these discrete capacitors is difficult because at the receiver, signal traces are often spaced 5 milli-inches (mils) apart while the most common capacitor size used in this application is 30 mils wide. Finally, placing discrete capacitors to implement mode compensation violates design-for-manufacturing

(DFM) rules pertaining to component-to-component spacing, which leads to higher assembly fallout and more expensive finished PCBs.

BRIEF DESCRIPTION OF THE DRAWINGS

The features and advantages of the present invention will become apparent from the following detailed description of the present invention in which:

Figure 1 is a diagram illustrating a parallel plate capacitor in which one
5 embodiment of the invention can be practiced.

Figure 2 is a diagram illustrating a computer-aided design (CAD) of buried intersignal capacitance mode compensation (BICMC) on a PCB according to one embodiment of the invention.

Figure 3 is a diagram of a partial view of signal paths in two adjacent conductive
10 signal layers of a PCB, illustrating an embodiment of the invention.

DETAILED DESCRIPTION

In the following description, for purposes of explanation, numerous details are set forth in order to provide a thorough understanding of the present invention.

However, it will be apparent to one skilled in the art that these specific details are not
5 required in order to practice the present invention.

While this invention has been described with reference to illustrative
embodiments, this description is not intended to be construed in a limiting sense.
Various modifications of the illustrative embodiments, as well as other embodiments of
the invention, which are apparent to persons skilled in the art to which the invention
10 pertains are deemed to lie within the spirit and scope of the invention.

Figure 1 is diagram illustrating a parallel plate capacitor 100 in which one
embodiment of the invention can be practiced.

The parallel plate capacitor 100 includes upper and lower plates, which form
upper conductive layer 101 and lower conductive layer 102, respectively. The upper and
15 lower conductive layers 101 and 102 are separated by a dielectric layer 103. The
dielectric layer 103 may be formed of materials such as a photoimagable polyimide and
epoxy resin. The dielectric layer 103 has a plurality of via holes (not shown), which
extend from the conductive layer 101 to the conductive layer 102. The via holes are
most commonly drilled and plated with metal, but could be fabricated in a variety of
20 ways. The via holes serve to electrically connect conductive paths on different
conductive layers of the PCB.

A parallel-plate capacitor is formed when the upper and lower conductive layers
101 and 102 are displaced with a particular proximity to one another such that the
voltage difference between the plates is proportional to the difference in the charges on

each plate. The capacitance of the parallel-plate capacitor 100 is a function of the area (A) of the plates formed on the upper and lower conductive layers 101 and 102, the distance (d) between the two conductive layers, and the permittivity (ϵ) of the dielectric layer separating the two conductive layers. In particular, the mutual capacitance (C) is
5 given by the following equation:

$$C = \epsilon A / d$$

In other words, the capacitance size is controlled by varying the distance between the two conductive layers 101 and 102 and/or the area of the plates on conductive layer 101 and/or conductive layer 102.

10 Figure 2 is an illustration of a top view computer-aided design (CAD) layout of the buried intersignal capacitance mode compensation (BICMC) on a printed circuit board (PCB) 200 according to one embodiment of the invention.

The PCB 200 is used for interconnecting integrated circuit (IC) chips and other electronic components and devices. The PCB 200 is formed by a substrate that supports
15 a plurality of insulated conductive trace layers (i.e., conductive layers 101 and 102). The insulated trace layers typically include surface (e.g., outer) conductive trace layers and embedded trace layers with selected trace layers connected as a ground plane and a power plane. Integrated circuits and electronic components and devices are mounted on an outer surface of the multiple layered printed circuit board and selectively connected
20 to the trace layers by plated-through holes called via holes or vias. As stated earlier, a via is an electrical conductor between the multiple layers in the PCB 200. The via may be made by drilling a via hole through the layers and plating the via hole with a conductive material.

Buried Intersignal Capacitance (BIC) is a method of arranging traces in a PCB that allows the board designer to create a specified amount of capacitance between signals. The BIC is formed by creating parallel plates on adjacent layers of circuit board while preserving the existing circuit board structure. The BIC is used in mode compensation to improve signal quality in the PCB. One of the effective placements of the mode compensating capacitor is between adjacent signal traces at the receiver end of the trace. Placing the mode compensation capacitor at the driver end of the two adjacent signal traces is an alternative way to improve signal quality in the PCB. Mode compensation counteracts the tendency of odd-mode crosstalk to travel faster than even-mode crosstalk through a microstrip transmission line. Even-mode crosstalk occurs when a signal line changes its level in one direction and an adjacent signal line changes its level in the same direction (i.e., the signal line is changed from low to high and the adjacent signal line is also changed from low to high). Odd-mode crosstalk occurs when a signal line changes its level in one direction and the adjacent signal line changes its level in the opposite direction (i.e., the signal line is changed from low to high and the adjacent signal line is changed from high to low).

The electric field associated with odd-mode crosstalk exists mostly in the air as it propagates down the microstrip transmission line. The electric field associated with even-mode crosstalk exists mostly in the insulating dielectric layer of the PCB as it propagates down the microstrip transmission line. Electric fields propagate faster through air than through the insulating dielectric layer of the PCB. This causes the difference in flight times between odd-mode and even-mode crosstalk, and can lead to signal quality issues at the receiver, especially from the time that odd-mode crosstalk arrives at the receiver end of a transmission line to the (later) time that even-mode crosstalk arrives at the receiver end of the transmission line. The function of the capacitance between the two adjacent signal lines is to match the speed of the odd-

mode to that of the even- mode. This greatly influences the unwanted crosstalk being transmitted down the signal lines, and also equalizes the arrival time of various data patterns representing even and odd modes. It is contemplated that the coupling of the capacitance between the two adjacent signal lines may also be used for other

5 applications such as to reduce jitter in a system clock.

In one embodiment, the PCB 200 includes multiple conductive layers (i.e., routing layers) and multiple intermediate dielectric layers (not shown). The multiple conductive layers include plurality signal planes, ground (GND) planes, and power (VCC) planes. The multiple conductive layers are separated by the multiple intermediate dielectric layers

10 (i.e., dielectric layer 103). The multiple conductive layers are interconnected by a plurality of vias 205₁ to 205_N, including connections to the power planes and connections to the ground planes, so that signal and power currents are transferred between the planes. The multiple conductor layers also contain metal floods that form plurality of upper and lower plates 211₁ to 211_M and 212₁ to 212_M, respectively, that

15 form a plurality of BICMC capacitors 210₁ to 210_M (where M is a positive integer). It is noted that in one embodiment, the metal flood may be made of copper.

The signal planes lie within the multiple conductive layers. The signal planes include a plurality of signal paths (i.e., traces) 215₁ to 215_P (where P is a positive integer) for conducting signals between integrated circuit chips. The ground planes or the power

20 planes also lie within the multiple conductive layers. These planes provide power supply to circuits connected to the multiple layers PCB structure and that current may be channeled from the power planes through the vias 205₁ to 205_N to supply power to the integrated circuit chips.

Metal floods such as copper, aluminum, gold, silver, lead/tin (solder),

25 molybdenum and others may also be formed in the power planes in a plurality of

isolated areas that form the plurality of lower plates 212₁ to 212_M. These lower plates are part of the parallel plates that form the plurality of capacitors 210₁ to 210_M.

Individual isolated areas 213₁ to 213_M are created around the plated holes or vias 205₁ to 205_M. The individual isolated areas are designed so that the lower plates and the

5 upper plates in the signal plane form the parallel plates that make up the plurality of capacitors 210₁ to 210_M.

The signal paths 215₁ to 215_P within the signal planes are connected to the metal flood 211₁ to 211_M within the same planes. Other signal paths within the signal planes are connected to the other planes in other layers by vias 205₁ to 205_N. The other layers
10 may be layers that contain ground planes and/or power planes. The signal, ground, and power planes may be on the outer surface layers or the inner layers.

In one embodiment, an outer or surface layer 255 and an inner layer (not shown) are separated by a dielectric layer. The inner layer contains a power plane that is mostly flooded with copper since it is a power delivery layer. The flooded copper areas are
15 divided into plurality of isolated smaller areas 213₁ to 213_M (shown in hatched since they are embedded beneath the surface layer 255), which are used to form a plurality of lower plates of the capacitors 210₁ to 210_M. The surface layer 255 contains a signal plane that has the plurality of signal paths 215₁ to 215_P. Each of the signal paths 215₁ to 215_P has a corresponding receiver end and a driver end. For example, a signal path 215₂ from the
20 signal plane, embedded in the surface layer 255, interconnects with the power plane, embedded in the inner layer, by a via 205₁. The via 205₁ is located inside the isolated area 213₁ (shown in hatched) that is formed in the power plane to create lower plate 212₁, one of the two parallel plates of the capacitor 210₁. The signal path 215₁, which is adjacent to the signal path 215₂, is connected to the copper flood on the signal plane to
25 form the upper plate of the parallel plates. The signal path 215₂ is connected the adjacent plane (i.e., power plane) through the via 205₁. The capacitor 210₁ is formed

between the two plates 211₁ and 212₁, and thus between the two signal paths 215₁ and 215₂. The signal paths 215₁ to 215_p may be adjacent to one another on the same signal plane or may be adjacent to one another on two different planes (i.e., the signal and power planes).

5 In one embodiment, capacitance is placed between adjacent signal paths at the receiver end of the signal paths in mode compensation. In other embodiments, the capacitance may be placed at different locations along the signal paths (i.e., the driver ends). The placing of capacitance between adjacent signal paths compensates for the different odd and even mode wave velocities through a microstrip. As stated earlier, the

10 mode compensation is used, in one embodiment, to help minimize the upward glitch seen on data lines in a system when one or more lines are held low and other data lines switch from low to high.

 Figure 3 is a diagram illustrating a partial view of a signal path in one signal layer and a partial view of the next plane layer adjacent to the signal layer of the PCB 200

15 according to one embodiment of the invention.

 A signal layer 301A includes signal paths 302A and 302B having vias 303A and 303B, respectively. The signal paths 302A and 302B carry signals between devices on the PCB 200. The two signal paths 302A and 302B may be horizontal adjacent to one another on the same plane (i.e., signal plane) or may be vertical adjacent to one another

20 on two different but adjacent planes. The via 303A as shown on the signal layer 301A is used to interconnect signal path 302A on the signal layer 301A to the signal path 302A on the next layer 301B. In other words, the signal path 302A extends down from the via 303A to beneath the signal path 302B, turns, and follows parallel to the signal path 302B.

A buried interconnect capacitance (as discuss in figure 2) is created between the signal path 302B of the signal layer 301A and the signal path 302A of the next layer 301B. These two paths run immediately above/below each other and are separated by a dielectric layers. The size of the capacitance can be controlled by adjusting the distance
5 between the signal path 302B on signal layer 301A and the signal path 302 on the adjacent routing layer 301B.

While this invention has been described with reference to illustrative embodiments, this description is not intended to be construed in a limiting sense. Various modifications of the illustrative embodiments, as well as other embodiments of
10 the invention, which are apparent to persons skilled in the art to which the invention pertains are deemed to lie within the spirit and scope of the invention.